

# Lecture 11

## Electromagnetic waves in Dielectric medium

# Electromagnetic waves in Dielectric medium

- In an isotropic dielectric medium, there are no fundamental charge carriers, hence the current density in a dielectric medium is zero ( $J=0$ ).
- There is no volume distribution of charge in the medium i.e. volume charge density is zero ( $\rho = 0$ ).
- Maxwell's eqs reduce to

$$\nabla \cdot \vec{E} = 0 \quad \dots(1) \quad D = \epsilon E$$

$$\nabla \cdot \vec{H} = 0 \quad \dots(2) \quad B = \mu H$$

$$\nabla \times \vec{E} = -\mu \frac{\partial H}{\partial t} \quad \dots(3)$$

$$\nabla \times \vec{H} = \epsilon \frac{\partial E}{\partial t} \quad \dots(4)$$

Taking the curl of eq 3 both sides

$$\nabla \times (\nabla \times \vec{E}) = \nabla \times \left( -\mu \frac{\partial \vec{H}}{\partial t} \right) = -\mu \frac{\partial}{\partial t} (\nabla \times \vec{H})$$

Now from vector identity

$$\nabla \times (\nabla \times \vec{E}) = \nabla(\nabla \cdot \vec{E}) - ((\nabla \cdot \nabla)\vec{E}) = \nabla(\nabla \cdot \vec{E}) - \nabla^2 \vec{E}$$

from Maxwell, s eq 4

$$\nabla \times \vec{H} = \varepsilon \frac{\partial \vec{E}}{\partial t}$$

$$\nabla(\nabla \cdot \vec{E}) - \nabla^2 \vec{E} = -\mu \frac{\partial}{\partial t} \left( \varepsilon \frac{\partial \vec{E}}{\partial t} \right)$$

from Maxwell, s eq 1

$$\nabla \cdot \vec{E} = 0$$

$$\nabla^2 \vec{E} = \mu \frac{\partial}{\partial t} \left( \varepsilon \frac{\partial \vec{E}}{\partial t} \right) = \mu \varepsilon \frac{\partial^2 \vec{E}}{\partial t^2}$$

$$\nabla^2 \vec{E} - \mu\epsilon \frac{\partial^2 \vec{E}}{\partial t^2} = 0$$

$$\nabla^2 \vec{E} - \frac{1}{v^2} \frac{\partial^2 \vec{E}}{\partial t^2} = 0 \quad \text{where } v = \frac{1}{\sqrt{\mu\epsilon}} \dots 5$$

Similarly Taking the curl of eq 4 both sides

$$\nabla \times (\nabla \times \vec{H}) = \nabla \times \left( \epsilon \frac{\partial \vec{E}}{\partial t} \right) = \epsilon \frac{\partial}{\partial t} (\nabla \times \vec{E})$$

But from Maxwell, s eq 3 and eq2

$$\nabla \times \vec{E} = -\mu \frac{\partial \vec{H}}{\partial t} \quad \text{and} \quad \nabla \cdot \vec{H} = 0$$

$$\nabla(\nabla \cdot \vec{H}) - \nabla^2 \vec{H} = \epsilon \frac{\partial}{\partial t} \left( -\mu \frac{\partial \vec{H}}{\partial t} \right)$$

$$0 - \nabla^2 H = \varepsilon \frac{\partial}{\partial t} \left( -\mu \frac{\partial H}{\partial t} \right)$$

$$\nabla^2 H = \mu \varepsilon \frac{\partial^2 H}{\partial t^2}$$

$$\nabla^2 H - \mu \varepsilon \frac{\partial^2 H}{\partial t^2} = 0$$

$$\nabla^2 H - \frac{1}{v^2} \frac{\partial^2 H}{\partial t^2} = 0 \quad \text{where } v = \frac{1}{\sqrt{\mu \varepsilon}} \quad \dots 6$$

Expression 5 and 6 are the wave equations for the propagation of electromagnetic waves in dielectric medium with a speed  $v$ .

$$v = \frac{1}{\sqrt{\mu \varepsilon}} = \frac{1}{\sqrt{\mu_0 \mu_r \varepsilon_0 \varepsilon_r}} = \frac{1}{\sqrt{\mu_r \varepsilon_r}} \times \frac{1}{\sqrt{\mu_0 \varepsilon_0}} = \frac{c}{\sqrt{\mu_r \varepsilon_r}} \quad \dots 7$$

Where  $\mu_r$  is the relative permeability and  $\varepsilon_r$  is the relative permittivity of the medium

since  $c = \frac{1}{\sqrt{\mu_0 \varepsilon_0}}$

as  $\mu_r > 1$  and  $\epsilon_r > 1$

$$v < c$$

Thus, the velocity of propagation of a wave in a dielectric medium is less than that in air or free space.

The refractive index of the dielectric medium is defined as

$$n = \frac{\text{speed of wave in vacuum}}{\text{speed of wave in medium}} = \frac{c}{v} = \sqrt{\mu_r \epsilon_r} \quad \dots 8$$

If the medium is non-magnetic then  $\mu_r = 1$

$$n = \sqrt{\epsilon_r} \quad \text{or} \quad n^2 = \epsilon_r$$

Thus the refractive index of a non-magnetic dielectric medium is equal to the square root of its relative permittivity.

# Solution of Electromagnetic waves for Dielectric medium

Assume we have a plane wave propagating in x direction (ie E, B not functions of y or z)

$$\nabla^2 E - \frac{1}{v^2} \frac{\partial^2 E}{\partial t^2} = 0 \quad \dots 9$$

$$\nabla^2 H - \frac{1}{v^2} \frac{\partial^2 H}{\partial t^2} = 0 \quad \dots 10$$

The wave solution of above eq in well known form may be written as

$$\vec{E}(r, t) = \vec{E}_0 e^{i(\vec{k} \cdot \vec{r} - \omega t)}$$

$$\vec{H}(r, t) = \vec{H}_0 e^{i(\vec{k} \cdot \vec{r} - \omega t)}$$

Where  $E_0$  and  $H_0$  are complex amplitudes which are constant in space and time

but  $\vec{k}$  is a wave propagation vector and defined as

$$\vec{k} = k \hat{n} = \frac{2\pi}{\lambda} \hat{n} = \frac{2\pi\nu}{v} \hat{n} = \frac{\omega}{v} \hat{n}$$

Where  $\hat{n}$  is a unit vector along the direction of propagation of em waves

The solution of plane em wave eqs in dielectric medium represented by eqs 9 and 10 satisfy Maxwell's equation 1 and 2 only when

$$\nabla \cdot \vec{E} = 0 \quad \text{when} \quad \vec{k} \cdot \vec{E} = 0$$

similarly

$$\nabla \cdot \vec{H} = 0 \quad \text{when} \quad \vec{k} \cdot \vec{H} = 0$$

This indicates that electric and magnetic fields are  $\perp$  to the direction of propagation vector  $k$  i.e the em waves in isotropic dielectric are transverse in nature.

Maxwell's Eq 3 and eq4 in isotropic dielectric

$$\nabla \times \vec{E} = -\mu \frac{\partial \vec{H}}{\partial t}$$

$$\nabla \times \vec{H} = \varepsilon \frac{\partial \vec{E}}{\partial t}$$



$$\vec{k} \times \vec{E} = \mu\omega\vec{H} \quad \dots 11$$

Similarly for  $\vec{k} \times \vec{H}$  may be obtained as

$$\vec{k} \times \vec{H} = -\varepsilon\omega\vec{E} \quad \dots 12$$

From eq 11, the field vector H is  $\perp$  to both k and E and according to eq 12, E is  $\perp$  to both k and H.

So its concluded that field vector E and H are mutually  $\perp$  to each other and also  $\perp$  to the direction of propagation of wave.

from eq 11

$$\mu\omega\vec{H} = \vec{k} \times \vec{E} \quad \dots A \quad (\vec{k} = k \hat{n})$$

$$\mu\omega\vec{H} = k \left( \hat{n} \times \vec{E} \right)$$

Where n is a unit vector along the direction of propagation of em waves

$$\vec{H} = \frac{k}{\mu\omega} \left( \hat{n} \times \vec{E} \right)$$

but  $\vec{k}$  is a wave propagation vector and defined as

$$\vec{k} = k \hat{n} = \frac{2\pi}{\lambda} \hat{n} = \frac{2\pi\nu}{v} \hat{n} = \frac{\omega}{v} \hat{n}$$

$$\therefore \vec{H} = \frac{1}{\mu v} \left( \hat{n} \times \vec{E} \right)$$

in terms of magnitude

$$|\vec{H}| = \sqrt{\frac{\epsilon}{\mu}} |\hat{n} \times \vec{E}| \quad \because k = \frac{\omega}{v} \quad \text{and } v = \frac{1}{\sqrt{\mu\epsilon}}$$

$$H = \sqrt{\frac{\epsilon}{\mu}} E$$

Now the ratio of magnitude of E to the magnitude of H is symbolized by Z

$$Z = \frac{E}{H} = \frac{E_0}{H_0} = \sqrt{\frac{\mu}{\epsilon}} = \sqrt{\frac{\mu_0 \mu_r}{\epsilon_0 \epsilon_r}}$$

This relation shows that the field vector E and H are in same phase.  
Z is called the wave impedance of isotropic dielectric medium.

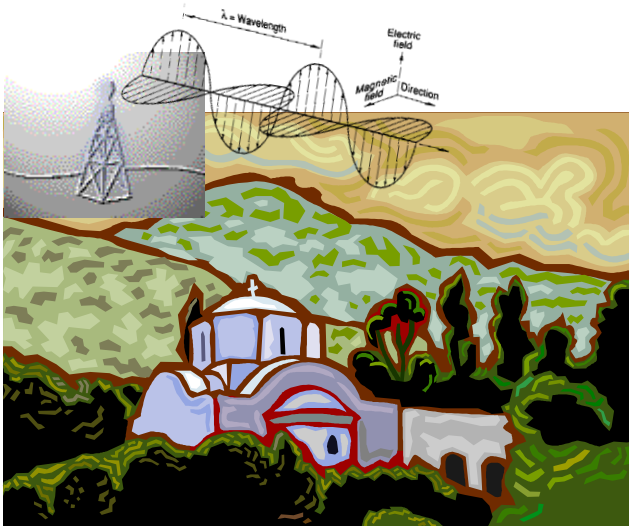
The wave impedance of medium is related to that of free space by the relation

$$Z = \sqrt{\frac{\mu_r}{\epsilon_r} \cdot \frac{\mu_0}{\epsilon_0}} = Z_0 \sqrt{\frac{\mu_r}{\epsilon_r}}$$

where  $Z_0 = \sqrt{\frac{\mu_0}{\epsilon_0}}$  is called the wave impedance  
in free space.

# Power in a wave

- A wave **carries power** and **transmits** it wherever it goes



The rate of flow of energy per unit area in wave is given by the *Poynting vector*.



# Poynting Vector Derivation

Taking scalar product of Maxwell 's eq 3 with H and eq 4 with E

$$H \cdot \nabla \times E = -H \cdot \frac{\partial B}{\partial t} \quad \dots 1$$

$$E \cdot (\nabla \times H) = E \cdot J + E \cdot \frac{\partial D}{\partial t} \quad \dots 2$$

Subtracting eq 2 from eq 1

$$H \cdot \nabla \times E - E \cdot (\nabla \times H) = -H \cdot \frac{\partial B}{\partial t} - E \cdot J - E \cdot \frac{\partial D}{\partial t}$$

$$H \cdot \nabla \times E - E \cdot (\nabla \times H) = - \left[ H \cdot \frac{\partial B}{\partial t} + E \cdot \frac{\partial D}{\partial t} \right] - E \cdot J \quad \dots 3$$

Now using vector identity

$$\nabla \cdot (A \times B) = B \cdot (\nabla \times A) - A \cdot (\nabla \times B) \text{ or in this case :}$$

$$\nabla \cdot (E \times H) = H \cdot (\nabla \times E) - E \cdot (\nabla \times H)$$

Eq 3 can be written as

$$\nabla \cdot (E \times H) = - \left[ H \cdot \frac{\partial B}{\partial t} + E \cdot \frac{\partial D}{\partial t} \right] - E \cdot J \quad \dots 4$$

Using relation  $B = \mu H$  and  $D = \epsilon E$ , in eq 4

$$\begin{aligned} \nabla \cdot (E \times H) &= - \left[ H \cdot \frac{\partial(\mu H)}{\partial t} + E \cdot \frac{\partial(\epsilon E)}{\partial t} \right] - E \cdot J \\ &= -\mu H \cdot \frac{\partial H}{\partial t} - \epsilon E \cdot \frac{\partial E}{\partial t} - E \cdot J \quad \dots 5 \end{aligned}$$

But

$$H \cdot \frac{\partial H}{\partial t} = \frac{1}{2} \frac{\partial(H)^2}{\partial t} \quad \text{and} \quad E \cdot \frac{\partial E}{\partial t} = \frac{1}{2} \frac{\partial(E)^2}{\partial t}$$

3/7/2013

$$\nabla \cdot (E \times H) = - \frac{\mu}{2} \frac{\partial(H)^2}{\partial t} - \frac{\epsilon}{2} \frac{\partial(E)^2}{\partial t} - E \cdot J$$

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$$\nabla \cdot (E \times H) = \frac{\partial}{\partial t} \left[ \frac{\mu H^2}{2} + \frac{\epsilon E^2}{2} \right] - E \cdot J \quad \dots 6$$

Now integrating eq 6 over a volume V bounded by surface S

$$\int_V \nabla \cdot (E \times H) dv = - \frac{\partial}{\partial t} \int_V \left( \frac{\epsilon}{2} E^2 + \frac{\mu}{2} H^2 \right) dv - \int_V E \cdot J dv \quad \dots 7$$

Using Gauss Divergence theorem  $\int_V \nabla \cdot (E \times H) dv = \oint_S (E \times H) ds$

Eq 7 becomes

$$\oint_S (E \times H) \cdot dS = - \frac{\partial}{\partial t} \int_V \left( \frac{\epsilon}{2} E^2 + \frac{\mu}{2} H^2 \right) dv - \int_V E \cdot J dv \quad \dots 9$$

Which means that the total power coming out of a volume is either due to the electric or magnetic field energy variations or is lost in ohmic losses.



# Poynting Vector

- Waves carry energy and information
- Poynting says that the net power flowing out of a given volume is = to the decrease in time in energy stored minus the conduction losses.

$$\vec{S} = \vec{E} \times \vec{H} \quad [\text{W/m}^2]$$

*Represents the instantaneous power density vector associated to the electromagnetic wave.*

- The Poynting vector has the same direction as the direction of propagation.

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# The Poynting Vector

Energy transport is defined by the Poynting vector  $\mathbf{S}$  as:

$$\vec{\mathbf{S}} \equiv \frac{\vec{\mathbf{E}} \times \vec{\mathbf{B}}}{\mu_0}$$

The direction of  $\mathbf{S}$  is the direction of propagation of the wave

$$S = \frac{EB}{\mu_0} = \frac{E^2}{\mu_0 c} = \frac{E^2}{Z_0} = \frac{E^2}{377\Omega}$$

# Electrostatic Boundary Conditions

- **At any point on the boundary,**
  - the components of  $\mathbf{E}_1$  and  $\mathbf{E}_2$  tangential to the boundary are equal

$$E_{1t} = E_{2t}$$

- the components of  $\mathbf{D}_1$  and  $\mathbf{D}_2$  normal to the boundary are equal

$$D_{1n} = D_{2n}$$

Hayt p-143

# Magnetic Boundary Conditions

- The normal component of  $\mathbf{B}_1$  and  $\mathbf{B}_2$  continuous across a interface:

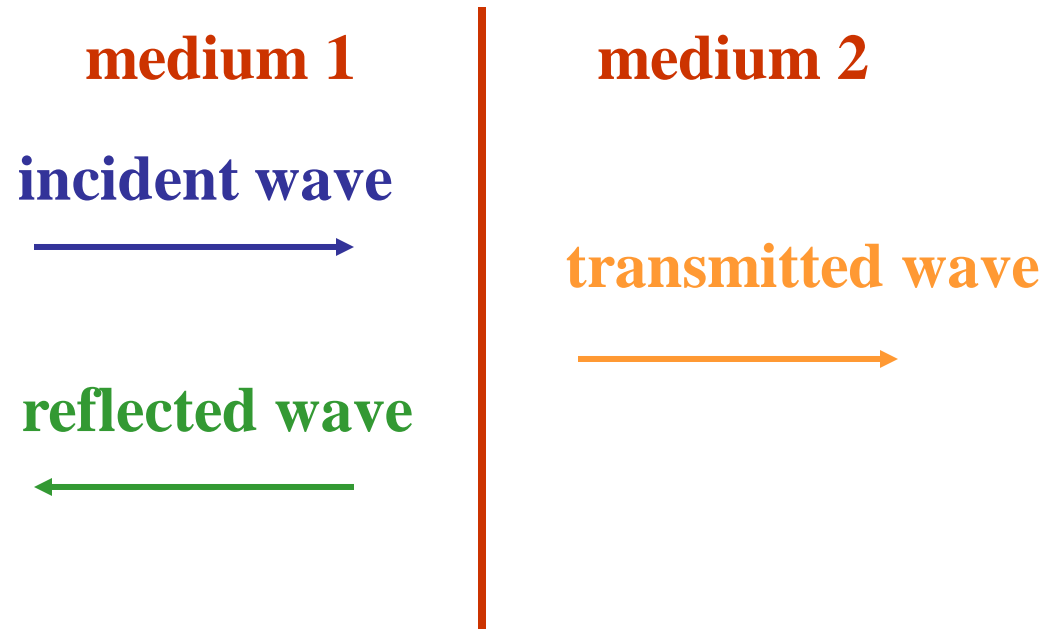
$$\boxed{B_{1n} = B_{2n}}$$

- The tangential component of a  $\mathbf{H}_1$  and  $\mathbf{H}_2$  to the boundary are equal

$$\boxed{H_{1t} = H_{2t}}$$

Hayt p-281

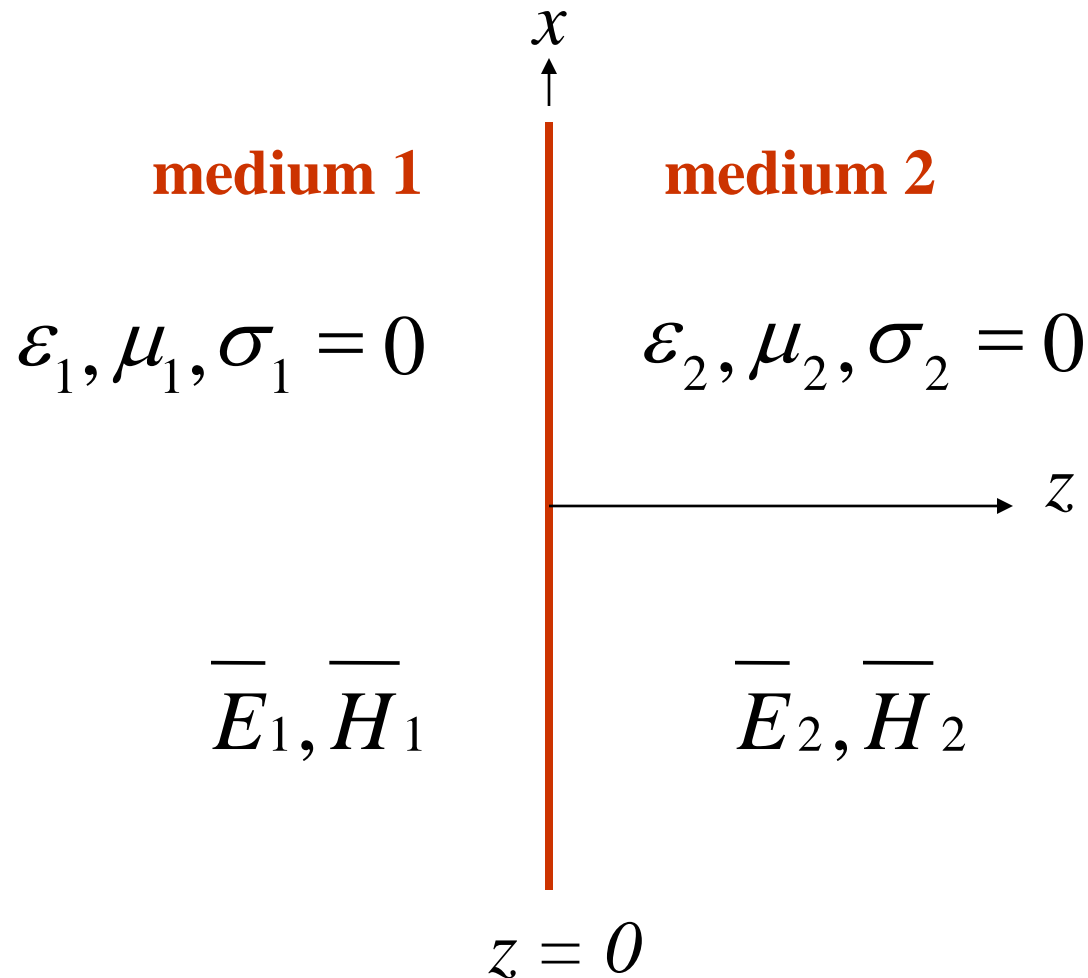
# Reflection and Transmission of Waves at Planar Interfaces



# Normal Incidence

- Consider both medium 1 and medium 2
- Let us place the boundary between the two media in the  $z = 0$  plane, and consider an incident plane wave which is traveling in the  $+z$ -direction.
- we assume that the electric field of the incident wave is in the  $x$ -direction.

# Normal Incidence





# Normal Incidence

- Incident wave

known

$$\overline{\mathbf{E}}_i = \hat{\mathbf{a}}_x E_{i0} e^{-j\beta_1 z}$$

$$\overline{\mathbf{H}}_i = \frac{1}{\eta_1} \hat{\mathbf{a}}_z \times \overline{\mathbf{E}}_i = \hat{\mathbf{a}}_y \frac{E_{i0}}{\eta_1} e^{-j\beta_1 z}$$

$$\beta_1 = \omega \sqrt{\epsilon_1 \mu_1} \qquad \eta_1 = \sqrt{\frac{\mu_1}{\epsilon_1}}$$

# Normal Incidence

- Reflected wave

unknown

$$\overline{E}_r = \hat{a}_x E_{r0} e^{+j\beta_1 z}$$

$$\overline{H}_r = \frac{1}{\eta_1} (-\hat{a}_z) \times \overline{E}_r = -\hat{a}_y \frac{E_{r0}}{\eta_1} e^{+j\beta_1 z}$$

# Normal Incidence

- Transmitted wave **unknown**

$$\overline{\mathbf{E}}_t = \hat{\mathbf{a}}_x E_{t0} e^{-j\beta_2 z}$$

$$\overline{\mathbf{H}}_t = \frac{1}{\eta_2} \hat{\mathbf{a}}_z \times \overline{\mathbf{E}}_t = \hat{\mathbf{a}}_y \frac{E_{t0}}{\eta_2} e^{-j\beta_2 z}$$

$$\beta_2 = \omega \sqrt{\epsilon_2 \mu_2} \quad \eta_2 = \sqrt{\frac{\mu_2}{\epsilon_2}}$$

# Normal Incidence

- The total electric and magnetic fields in medium 1 are

$$\overline{\mathbf{E}}_1 = \overline{\mathbf{E}}_i + \overline{\mathbf{E}}_r = \hat{\mathbf{a}}_x \left[ E_{i0} e^{-j\beta_1 z} + E_{r0} e^{+j\beta_1 z} \right]$$
$$\overline{\mathbf{H}}_1 = \overline{\mathbf{H}}_i + \overline{\mathbf{H}}_r = \hat{\mathbf{a}}_y \left[ \frac{E_{i0}}{\eta_1} e^{-j\beta_1 z} - \frac{E_{r0}}{\eta_1} e^{+j\beta_1 z} \right]$$

# Normal Incidence

- The total electric and magnetic fields in medium 2 are

$$\overline{E}_2 = \overline{E}_t = \hat{a}_x E_{t0} e^{-j\beta_2 z}$$

$$\overline{H}_2 = \overline{H}_t = \hat{a}_y \frac{E_{t0}}{\eta_2} e^{-j\beta_2 z}$$

# Normal Incidence

- To determine the unknowns  $E_{r0}$  and  $E_{t0}$ , we must enforce the BCs at  $z = 0$ :

$$\overline{E}_1(z = 0) = \overline{E}_2(z = 0)$$
$$\overline{H}_1(z = 0) = \overline{H}_2(z = 0)$$

# Normal Incidence)

- From the BCs we have

$$E_{i0} + E_{r0} = E_{t0}$$

$$\frac{E_{i0}}{\eta_1} - \frac{E_{r0}}{\eta_1} = \frac{E_{t0}}{\eta_2}$$

or

$$\eta_1 \quad \eta_1 \quad \eta_2$$

$$E_{r0} = \frac{\eta_2 - \eta_1}{\eta_2 + \eta_1} E_{i0}, \quad E_{t0} = \frac{2\eta_2}{\eta_2 + \eta_1} E_{i0}$$

# Reflection and Transmission Coefficients

- Define the *reflection coefficient* as

$$\Gamma = \frac{E_{r0}}{E_{i0}} = \frac{\eta_2 - \eta_1}{\eta_2 + \eta_1}$$

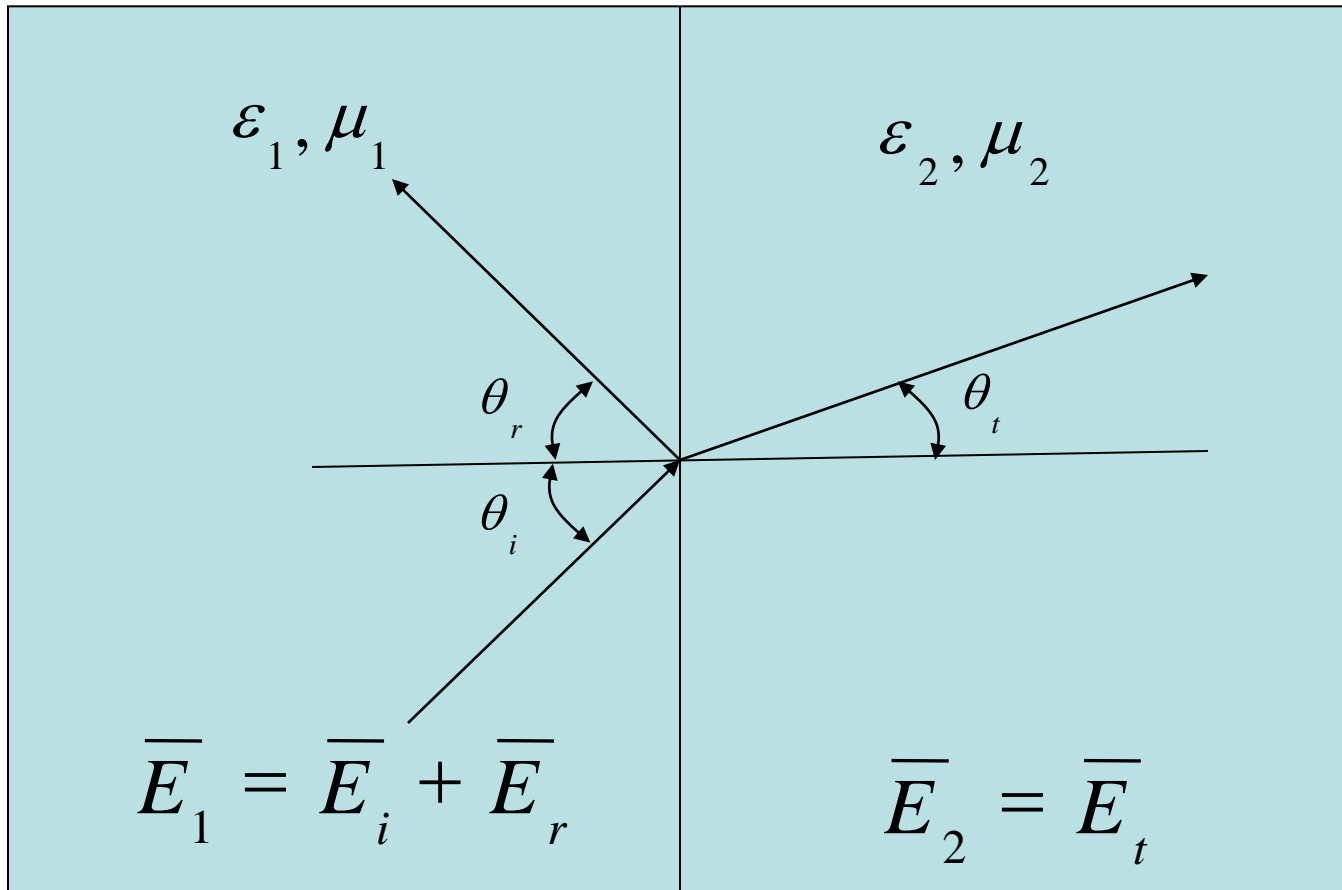
- Define the *transmission coefficient* as

$$\tau = \frac{E_{t0}}{E_{i0}} = \frac{2\eta_2}{\eta_2 + \eta_1}$$

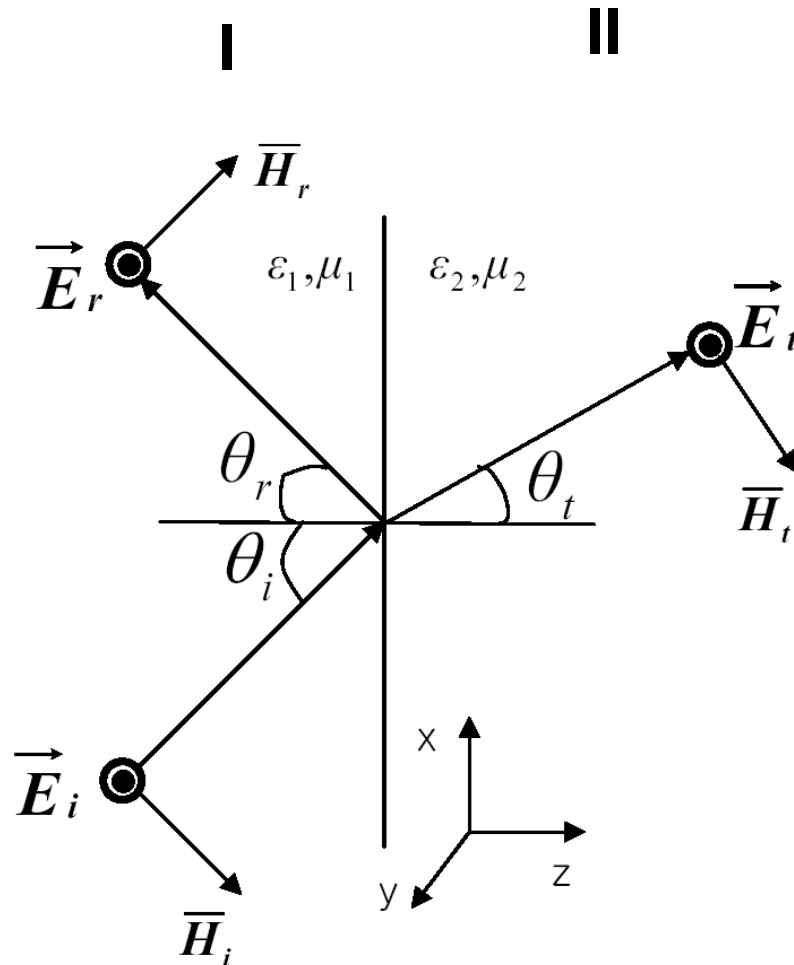


# Oblique Incidence

$$z = 0$$



# Perpendicular Polarization



$$\vec{E}_i = E_{i0} \exp(-j\beta_1 z) a_x$$

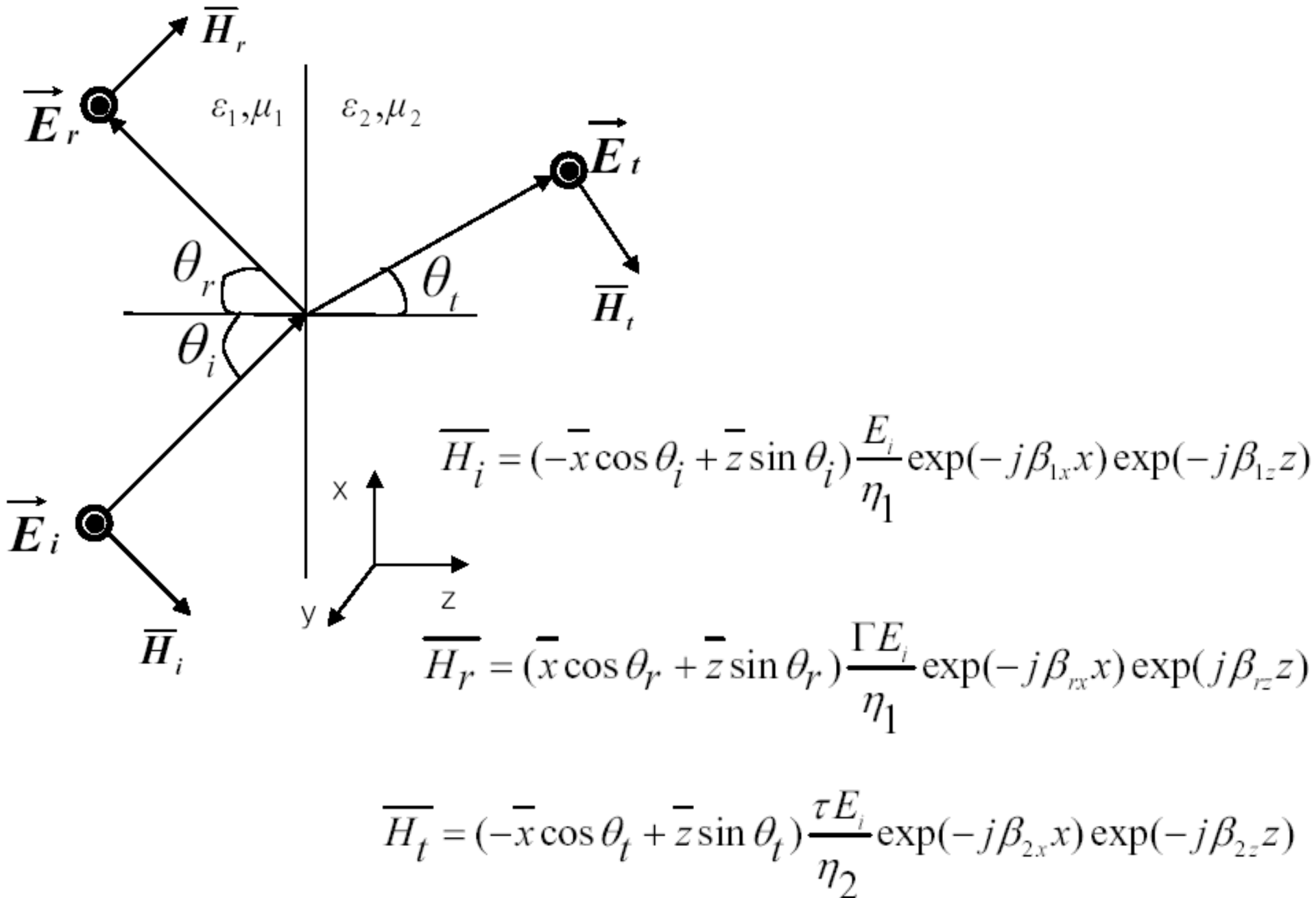
$$\vec{E}_i = \vec{y} E_i \exp(-j\beta_{1x} x) \exp(-j\beta_{1z} z)$$

$$\vec{E}_r = \vec{y} \Gamma E_i \exp(-j\beta_{rx} x) \exp(j\beta_{rz} z)$$

$$\vec{E}_t = \vec{y} \tau E_i \exp(-j\beta_{2x} x) \exp(-j\beta_{2z} z)$$

$$\Gamma = \frac{E_r}{E_i}$$

$$\tau = \frac{E_t}{E_i}$$



B.C.'s at  $z=0$

1)  $\bar{E}_{\text{tan}}$  continuous ( $\bar{E}_i + \bar{E}_r = \bar{E}_t$ )

$$\bar{E}_i = \bar{y} E_i \exp(-j\beta_{1x}x) \exp(-j\beta_{1z}z)$$

$$\bar{E}_r = \bar{y} \Gamma E_i \exp(-j\beta_{rx}x) \exp(j\beta_{rz}z)$$

$$\bar{E}_t = \bar{y} \tau E_i \exp(-j\beta_{2x}x) \exp(-j\beta_{2z}z)$$

$$\exp(-j\beta_{1x}x) + \Gamma \exp(-j\beta_{rx}x) = \tau \exp(-j\beta_{2x}x) \quad \dots\dots(\mathbf{A})$$

$$\beta_{1x} = \beta_{rx} = \beta_{2x}$$

$$1 + \Gamma = \tau \quad \dots\dots(\mathbf{B})$$

B.C.'s at  $z=0$

2)  $\bar{H}_{\text{tan}}$  continuous ( $\bar{H}_i + \bar{H}_r = \bar{H}_t$ )

$$-\frac{\cos \theta_i}{\eta_1} + \frac{\cos \theta_i}{\eta_1} \Gamma = -\frac{\cos \theta_t}{\eta_2} \tau \quad \dots\dots(\text{C})$$

With  $1 + \Gamma = \tau$ ,

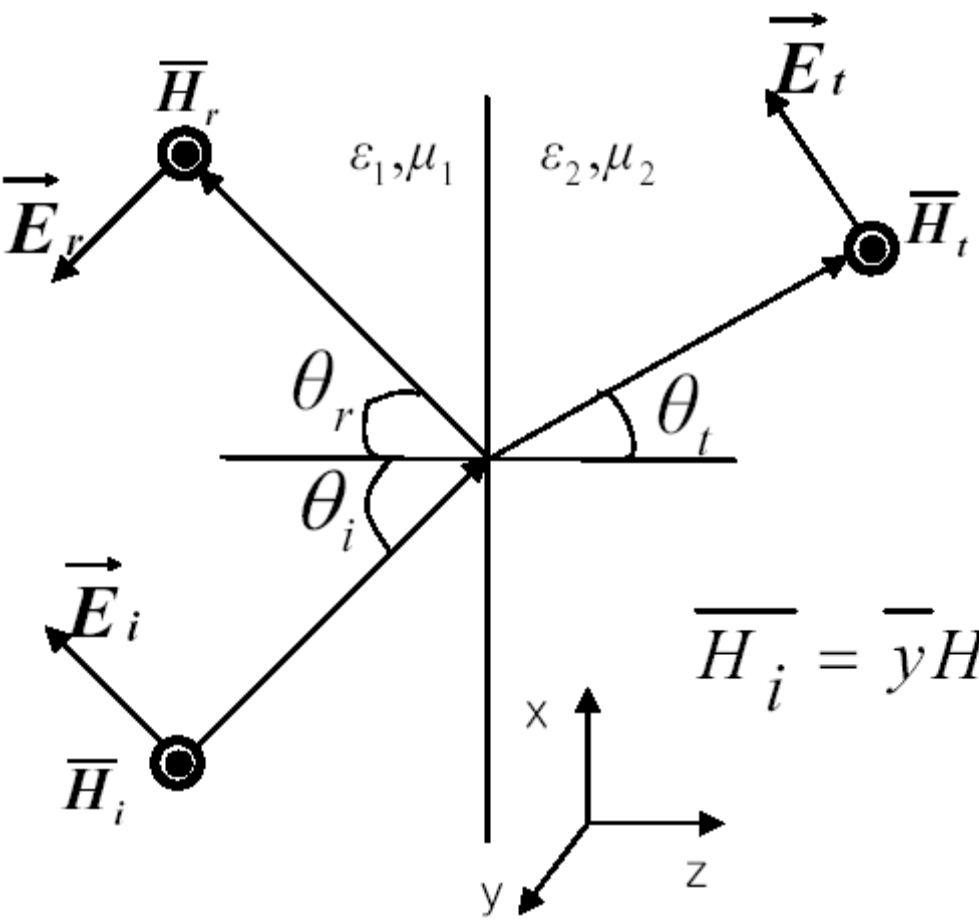
**reflection coefficient**

$$\Gamma_{\perp} = \frac{\eta_2 / \cos \theta_t - \eta_1 / \cos \theta_i}{\eta_2 / \cos \theta_t + \eta_1 / \cos \theta_i}$$

**transmission coefficient**

$$\tau_{\perp} = \frac{2\eta_2 / \cos \theta_t}{\eta_2 / \cos \theta_t + \eta_1 / \cos \theta_i}$$

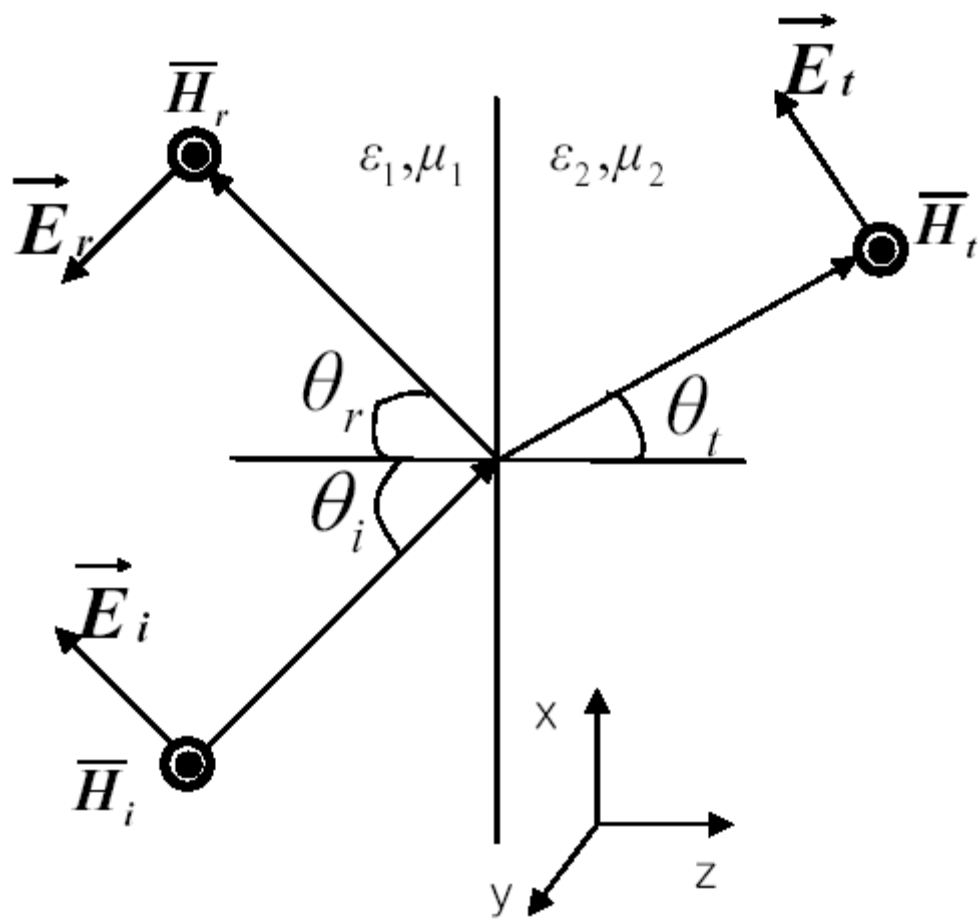
# Parallel Polarization



$$\overline{H}_i = \overline{y} H_i \exp(-j\beta_{1x}x) \exp(-j\beta_{1z}z)$$

$$\overline{H}_r = \overline{y} H_r \exp(-j\beta_{rx}x) \exp(j\beta_{rz}z)$$

$$\overline{H}_t = \overline{y} H_t \exp(-j\beta_{2x}x) \exp(-j\beta_{2z}z)$$



$$\mathbf{E} = \mathbf{H}\eta$$

$$\overline{E}_i = (\overline{x} \cos \theta_i - \overline{z} \sin \theta_i) H_i \eta_1 \exp(-j\beta_{1x}x) \exp(-j\beta_{1z}z)$$

$$\overline{E}_r = (-\overline{x} \cos \theta_r - \overline{z} \sin \theta_r) H_r \eta_1 \exp(-j\beta_{rx}x) \exp(j\beta_{rz}z)$$

$$\overline{E}_t = (\overline{x} \cos \theta_t - \overline{z} \sin \theta_t) H_t \eta_2 \exp(-j\beta_{2x}x) \exp(-j\beta_{2z}z)$$

$$\frac{H_r}{H_i} = -\Gamma \quad \frac{H_t}{H_i} = \frac{E_t / \eta_2}{E_i / \eta_1} = \tau \frac{\eta_1}{\eta_2}$$

B.C.'s at  $z=0$

1)  $\bar{H}_{\text{tan}}$  continuous ( $\bar{H}_i + \bar{H}_r = \bar{H}_t$ )

$$\bar{H}_i = \bar{y} H_i \exp(-j\beta_{1x}x) \exp(-j\beta_{1z}z)$$

$$\bar{H}_r = \bar{y} H_r \exp(-j\beta_{rx}x) \exp(j\beta_{rz}z)$$

$$\bar{H}_t = \bar{y} H_t \exp(-j\beta_{2x}x) \exp(-j\beta_{2z}z)$$



$$\exp(-j\beta_{1x}x) - \Gamma \exp(-j\beta_{rx}x) = \frac{\eta_1}{\eta_2} \tau \exp(-j\beta_{2x}x)$$

$$\beta_{1x} = \beta_{rx} = \beta_{2x}$$

$$1 - \Gamma = \frac{\eta_1}{\eta_2} \tau \quad \dots\dots(\mathbf{A})$$

B.C.'s at  $z=0$

2)  $\bar{E}_{\tan}$  continuous

$$\cos \theta_i + \Gamma \cos \theta_i = \tau \cos \theta_t \quad \dots\dots(\mathbf{B})$$

## *reflection coefficient*

$$\Gamma_{11} = \frac{\eta_2 \cos \theta_t - \eta_1 \cos \theta_i}{\eta_2 \cos \theta_t + \eta_1 \cos \theta_i}$$

## *transmission coefficient*

$$\tau_{11} = \frac{2\eta_2 \cos \theta_i}{\eta_2 \cos \theta_t + \eta_1 \cos \theta_i}$$